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LIQUID DRYSTAL DISPLAY DEVICE

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Title : LIQUID CRYSTAL DISPLAY DEVICE

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CROSS-REFERENCE TO RELATED APPLICATIONS

This application is based upon and claims the benefit of priority from the prior Japanese Patent Application No. 2003-095287, filed on March 31, 2003, the entire contents of which are incorporated herein by reference.

BACKGROUND OF THE INVENTION

[Field of the Invention]

The present invention relates to a liquid crystal display device, and more particularly, to changeover control of a driving power supply in the liquid crystal display device.

[Description of the Related Art]

In recent years, liquid crystal display devices are in wide use in devices for mobile use such as portable devices. In accordance therewith, a demand for low power consumption liquid crystal display devices suitable for outdoor use is increasing and reflective liquid crystal display devices are drawing attention.

The reflective liquid crystal display device is suitable for reducing power consumption and for outdoor use since it constantly utilizes external light as a light source without using any backlight. However, since the reflective liquid crystal display device uses external light as a light source, especially in a liquid crystal display device of an

active matrix type, an afterimage remains visible by a user due to residual charges in liquid crystal whenever the power supply is cut off (power-off), which results in deteriorated display quality.

A method of quickly clearing the display at the power-off time in the conventional reflective liquid display device is disclosed in the patent document 1 (Japanese Patent Application Laid-open No. Hei 1-170986). A reflective liquid crystal display device disclosed in the patent document 1 is configured such that a driving power supply is directly fed from a power source of the display device to a source driving circuit for driving source lines (source signal lines) of a display part, and a driving power supply is fed to a gate driving circuit for driving gate lines (gate signal lines) from the power source via a power holding circuit having a capacity large enough to hold a power for a prescribed time.

Then, a power-off signal generated based on the detection of a power-off state is inputted to a gate driving circuit, so that the gate lines connected to the gate driving circuit are all activated (activated to turn on transistors connected to the gate lines) simultaneously using the power held in the power holding circuit. Consequently, residual charges in liquid crystal at the power-off time are discharged in a short time to clear the display, so that an afterimage is prevented from being visible.

Further, another example of the conventional art is disclosed in, for example, the patent document 2 (Japanese Patent Application Laid-open No. 2001-195025).

However, if a driving waveform of gate lines at a normal driving time is a rectangular wave as in the conventional reflective liquid crystal display device described above, unevenness in luminance inclination may possibly occur in a display part between pixels closer to the gate driving circuit and pixels distant therefrom. Methods of solving this unevenness in luminance inclination include a method in which a voltage of a driving power supply (hereinafter, referred to as a "gate-on power supply") V_{gon} for activating gate lines is varied with time (in a pulsed manner) to be fed to the gate driving circuit, thereby blunting gate driving waveforms V_{OUT1} , V_{OUT2} , ... as shown in Fig. 5 (see, for example, the patent document 3 (Japanese Patent Application Laid-open No. 2001-125069)).

The use of such gate driving waveforms V_{OUT1} , V_{OUT2} ... whose falling edges are blunted can reduce luminance unevenness in the direction in which the gate lines extend. Note that V_{off} is a driving power supply for inactivating the gate lines (hereinafter, referred to as a "gate-off power supply") in Fig. 5. Further, a circuit for generating the gate-on power

supply V_{gon} as shown in Fig. 5 will be referred to as a luminance inclination circuit.

However, the method of reducing luminance inclination unevenness using the luminance inclination circuit has a problem that the power holding circuit that causes residual charges in liquid crystal to be quickly discharged to clear the display at the power-off time as described above is not applicable to the gate driving circuit. In other words, though display quality at a normal driving time is improved, display quality deterioration due to the afterimage at the power-off time cannot be avoided. This is because in the method of reducing luminance inclination unevenness, the voltage of the gate-on power supply V_{gon} is varied with time at the normal driving time, and if the power holding circuit is used, voltage fluctuation is made small due to its large capacity or the like, which makes it difficult to cause voltage variation (makes it difficult to generate a blunted waveform).

SUMMARY OF THE INVENTION

An object of the present invention is to realize improved display quality in a liquid crystal display device not only at a normal driving time but also at a power-off time.

A liquid crystal display device of the present invention is characterized in that at least one

internal power supply is automatically changed from a first power supply to a second power supply when an input power supply fed to the liquid crystal display device is cut off.

According to the present invention, the automatic changeover of the internal power supply is made possible in such a manner that the first power supply shown by the power supply Vgon in Fig. 5 is used as the internal power supply when the input power supply is fed, and the second power supply by a held power is used as the internal power supply when the input power supply is cut off. Consequently, a good quality display image with reduced unevenness in luminance inclination can be displayed at the normal driving time during which the input power supply is fed. In addition, when the input power supply is cut off, residual charges in liquid crystal are discharged in a short time, so that an afterimage can be prevented from persisting.

BRIEF DESCRIPTION OF THE DRAWINGS

Fig. 1 is a block diagram showing characteristic elements of a liquid crystal display device according to an embodiment;

Fig. 2A and Fig. 2B are diagrams showing circuit configuration examples of the characteristic elements of the liquid crystal display device according to this embodiment;

Fig. 3 is a block diagram showing a configuration example of the liquid crystal display device of this embodiment;

Fig. 4 is a timing chart showing examples of operational waveforms of the liquid crystal display device according to this embodiment; and

Fig. 5 is a chart showing an example of conventional gate driving waveforms.

DETAILED DESCRIPTION OF THE PREFERRED EMBODIMENTS

Hereinafter, an embodiment of the present invention will be explained based on the drawings.

Fig. 1 is a block diagram showing characteristic elements of a liquid crystal display device according to an embodiment of the present invention. It should be noted that the liquid crystal display device to be explained below is not limited to a specific type, but it is suitable for use as a liquid crystal display device which sometimes utilizes external light as a light source, for example, a reflective liquid crystal display device and a transflective liquid crystal display device (a liquid crystal display device that can be a reflection type and a transmission type according to outside brightness, control, and so on).

In Fig. 1, a luminance inclination circuit 1 changes a voltage of a gate-on power supply V_{on} for activating gate lines, in synchronization with an

output of a gate driving circuit 4, and outputs a gate-on power supply for gate driving circuit as shown by the power supply V_{gon} in Fig. 5. The use of this gate-on power supply for gate driving circuit makes it possible to intentionally blunt falling edges of outputs (gate driving voltages) $VOUT_i$ (i is a suffix and a natural number: $i = 1$ to n) of the gate driving circuit 4. This can reduce luminance unevenness in a direction in which the gate lines extend.

A power holding circuit 2 holds a power fed by the gate-on power supply V_{on} for a predetermined period of time and it is constituted using capacitors or the like having a sufficiently large capacity.

A power supply changeover circuit 3 selectively outputs one of an output voltage of the luminance inclination circuit 1 and an output voltage of the power holding circuit 2 to the gate driving circuit 4 according to the voltage level of an input power supply (device power supply) V_{in} fed from a not-shown power supply device or the like. Specifically, when a voltage value of the device power supply V_{in} is higher than a predetermined voltage value, the power supply changeover circuit 3 outputs the output voltage of the luminance inclination circuit 1 to the gate driving circuit 4, while, when the voltage value of the device power V_{in} is equal to or lower than the predetermined voltage value, it outputs the output

voltage of the power holding circuit 2 to the gate driving circuit 4.

The gate driving circuit 4 has a gate-on power supply terminal Vgon, an all-output-on terminal Vxon, a gate-off power supply terminal Voff, and a power supply terminal Vdd for driving each logic, and these terminals are supplied with the output voltage of the power supply changeover circuit 3, the device power supply Vin, a gate-off power supply Voff, and a logic power supply Vcc respectively.

When the device power supply Vin fed to the all-output-on terminal Vxon is high level, the gate driving circuit 4 outputs pulse signals in sequence every one gate line period from outputs VOUT1 to VOUTn based on not-shown inputted control signals such as a gate start pulse and a gate shift clock. Therefore, the gate driving circuit 4 (shifts and) drives the plural gate lines provided in a later-described display part in sequence starting from an upper area of a screen. By this operation, transistors (TFTs; thin film transistors) connected to the selected gate lines are turned on, and display data (tone voltages) are written to respective pixels associated with the selected gate lines. Note that the pulse signals are generated by synthesizing the output of the luminance inclination circuit 1 and the gate-off power supply Voff.

On the other hand, when the device power supply V_{in} fed to the all-output-on terminal V_{xon} turns to low level, the gate driving circuit 4 asynchronously outputs a voltage, which is supplied to the gate-on power supply terminal V_{gon} , to all the outputs V_{OUT1} , V_{OUT2} , ..., V_{OUTn} of the gate driving circuit 4 irrespective of other input signals.

Fig. 2A and Fig. 2B are diagrams showing circuit configuration examples of the characteristic elements of the liquid crystal display device according to this embodiment.

Fig. 2A is a diagram showing a concrete circuit configuration of the luminance inclination circuit 1, the power holding circuit 2, and the power supply changeover circuit 3 shown in Fig. 1.

As shown in Fig. 2A, the power holding circuit 2 is constituted of two capacitors $C1$, $C2$ in which electrodes on one side thereof are connected to the gate-on power supply V_{on} and the other electrodes are connected to the ground (grounded). Incidentally, Fig. 2A shows as an example the case where the power holding circuit 2 is constituted of the two capacitors, but it only needs to have a sufficient capacity, and the number of capacitors is arbitrary.

The power supply changeover circuit 3 is constituted of four resistances $R1$ to $R4$, two n-channel transistors (FETs: field effect transistors) $NT1$, $NT2$, and one p-channel transistor (FET) $PT1$.

The resistances R1, R2 connected in series are connected in series between the aforesaid electrodes on one side of the capacitors C1, C2 and the ground. Here, resistance values of the resistances R1, R2 are determined appropriately so that the transistor NT1 can be turned on when the device power supply Vin is off (when the device power supply Vin is cut off).

A drain of the transistor NT2 is connected to a mutual connection point (a node NA) between the resistances R1 and R2, a source thereof is connected to the ground, and a gate thereof is fed with the device power supply Vin.

Further, the resistances R3, R4 connected in series are connected in series between the aforesaid electrodes on one side of the capacitors C1, C2 and a drain of the transistor NT1. A gate of the transistor NT1 is connected to the node NA and a source thereof is connected to the ground.

A source of the transistor PT1 is connected to the aforesaid electrodes on one side of the capacitors C1, C2, a drain thereof is connected to the gate-on power supply terminal Vgon of the gate driving circuit 4, and a gate thereof is connected to a mutual connection point (a node NB) between the resistances R3 and R4.

The luminance inclination circuit 1 is constituted of three resistances R5, R6, R7, two n-channel transistors (FETs) NT3, NT4, and one p-

channel transistor (FET) PT2. Further, since a voltage fluctuation in synchronization with the outputs VOUTi of the gate driving circuit 4 is required, a gate clock signal GCLK for shifting the outputs VOUTi of the gate driving circuit 4 and a luminance inclination circuit control signal XGCLK that is inverted from the gate clock signal GCLK are inputted to the luminance inclination circuit 1. Incidentally, the gate clock signal GCLK and the luminance inclination circuit control signal XGCLK are generated in a later-described timing generating circuit.

The resistances R5, R6 connected in series are connected between the source of the transistor PT1 and a drain of the transistor NT3. A source of the transistor NT3 is connected to the ground and the gate clock signal GCLK is inputted to a gate thereof.

A source of the transistor PT2 is connected to the source of the transistor PT1, a drain thereof is connected to the gate-on power supply terminal Vgon, and a gate thereof is connected to a mutual connection point (a node NC) between the resistances R5 and R6. A drain of the transistor NT4 is connected to an end of the resistance R7 whose other end is connected to the gate-on power supply terminal Vgon, a source thereof is connected to the ground, and the luminance inclination circuit control signal XGCLK is inputted to a gate thereof.

With the above-described configuration, when the voltage value of the device power supply V_{in} is higher than the predetermined voltage value (high level), the potential of the node NA turns to low level (0(zero) V), so that the node NB turns to high level to turn off the transistor PT1. Consequently, the output voltage of the luminance inclination circuit 1 is supplied to the gate-on power supply terminal Vgon.

On the other hand, when the voltage value of the device power supply V_{in} is equal to or lower than the predetermined voltage value (low level), the potential of the node NA turns to high level, so that the node NB turns to low level to turn on the transistor PT1. Further, the gate clock signal GCLK and the luminance inclination circuit control signal XGCLK are shut off (suspended) in accordance with the decrease in the voltage value of the device power supply V_{in} . Consequently, the output voltage of the power holding circuit 2 is supplied to the gate-on power supply terminal Vgon.

Fig. 2B is a diagram showing a concrete circuit configuration of a power supply circuit 7-1 for supplying the logic power supply V_{cc} to the gate driving circuit 4. The power supply circuit 7-1 is provided in a later-described internal power supply generating circuit. The power supply circuit 7-1 is constituted of two capacitors C3, C4 and a diode 31.

An anode of the diode 31 is connected to the device power supply V_{in} and a cathode thereof is connected to the power supply terminal V_{dd} . The capacitors C3, C4 are connected in parallel between a mutual connection point of the cathode of the diode 31 and the power supply terminal V_{dd} and the ground. Incidentally, the case where the power supply circuit 7-1 is constituted of the two capacitors is shown as an example, but the number of capacitors is arbitrary. Further, noise filtering capacitors that are generally disposed in large number in power supply lines may be used as the capacitors C3, C4 to supply a power by residual charges thereof, and this will suffice for driving.

Fig. 3 is a block diagram showing a configuration example of the liquid crystal display device according to this embodiment. Note that, in Fig. 3, the same reference numerals and symbols are used to designate blocks and so on having the same functions as those of the blocks and so on shown in Fig. 1, and repeated explanation will be omitted.

In Fig. 3, a source driving circuit 5 supplies a voltage according to the tone level to a plurality of source lines provided in a display part 6. Specifically, based on a control signal SCTL such as a tone signal (display data signal) inputted from a timing generating circuit 8, the source driving circuit 5 selects a voltage according to the tone

signal from voltages LV at respective tone levels generated in a tone voltage generating circuit 9, and outputs all outputs thereof in synchronization with the outputs VOUTi of the gate driving circuit 4. This makes it possible to supply (write) the tone voltages to the respective pixels associated with the gate lines via the thin film transistors connected to the gate lines selected in the display part 6.

In the display part 6, a plurality of gate lines and a plurality of source lines are arranged in matrix, and pixels for displaying an image are arranged at intersections of the gate lines and the source lines. Each of the pixels includes a thin film transistor whose gate and source are connected to the gate line and the source line.

The internal power supply generating circuit 7 includes the power supply circuit 7-1 shown in Fig. 2B, and it generates, from the device power supply Vin, power supplies at voltages (for example, 24 V, -5.5 V, and so on) used in the respective circuits. The power supplies generated in the internal power supply generating circuit 7 include the gate-on power supply Von, the gate-off power supply Voff, the logic power supply Vcc, a reference power supply Vref for generating a voltage according to the tone level, and so on.

Based on an image input signal SIG inputted from an external part, the timing generating circuit 8

generates a control signal GCTL (for example, the gate start pulse, the gate shift clock signal, the luminance inclination circuit control signal, and so on) for the gate driving circuit 4 and the control signal SCTL (the clock signal and so on) for the source driving circuit 5. Further, the timing generating circuit 8 outputs the generated control signal GCTL to the gate driving circuit 4 and synchronously outputs the generated control signal SCTL and tone signals of respective RGB colors to the source driving circuit 5.

Using the fed reference power supply V_{ref} , the tone voltage generating circuit 9 generates the voltages at the respective tone levels that are to be supplied to liquid crystal via the thin film transistors arranged in the display part 6.

Next, the operation of the liquid crystal display device according to this embodiment will be explained in detail based on Fig. 4. Fig. 4 is a timing chart showing examples of operational waveforms of the liquid crystal display device according to this embodiment. Note that only the driving of the gate lines will be explained below, referring to Fig. 2 when necessary, but the driving of the source lines is the same as that of the conventional liquid crystal display device.

First, at a normal driving time during which the device power supply V_{in} is normally fed to the liquid

crystal display device (when the device power supply V_{in} is 3.3 V), the transistor NT2 whose gate is fed with the device power supply V_{in} is on (in a continuity state). Accordingly, the potential of the node NA turns to 0(zero) V and the transistor NT1 turns to an off state (non-continuity state). In accordance therewith, the potential of the node NB becomes equal to the potential of the gate-on power supply V_{on} , and the transistor PT1 whose gate is connected to the node NB turns off since the potentials supplied to the source and gate thereof become equal to each other.

Further, in the luminance inclination circuit 1, to which the gate clock signal GCLK and the luminance inclination circuit control signal XGCLK are inputted as described above, the transistor NT3 turns on when the gate clock signal GCLK is 3.3 V. Accordingly, the potential of the node NC becomes about 20 V and the transistor PT1 turns on, so that the gate-on power supply V_{on} is outputted to the gate-on power supply terminal V_{gon} of the gate driving circuit 4. At this time, since the luminance inclination circuit control signal XGCLK is 0(zero) V, the transistor NT4 is off, so that no influence is given to the output of the transistor PT2.

Next, when the gate clock signal GCLK is 0(zero) V, the transistor NT3 turns off. Accordingly, the potential of the node NC becomes equal to that of the

gate-on power supply V_{on} , and the potentials of the gate and source of the transistor PT2 become equal to each other, so that the transistor PT2 turns off. At this time, since the luminance inclination circuit control signal XGCLK is 3.3 V, the transistor NT4 turns on, and the potential of the gate-on power supply terminal V_{gon} is lowered toward 0(zero) V side.

Thus, at the normal driving time, the output voltage of the luminance inclination circuit 1, the respective power supply voltages such as the gate-off power supply V_{off} , and the control signal GCTL are inputted to the gate driving circuit 4, so that the outputs V_{OUT1} , V_{OUT2} , ..., V_{OUTn} of the gate driving circuit 4 as shown in Fig. 4 are sequentially outputted every one gate line period. The voltage supplied to the gate-on power supply terminal V_{gon} is changed in synchronization with falling edges of the gate clock signal GCLK, so that each of the outputs V_{OUTi} of the gate driving circuit 4 also has a driving waveform whose falling edges are intentionally blunted. Therefore, luminance unevenness in the direction in which the gate lines extend in the display part 6 can be reduced, which makes it possible to obtain a good quality display image.

At the power-off time when the device power supply V_{in} fed to the liquid crystal display device is cut off (at a time t_1 at which the device power

supply V_{in} turns to 0 (zero) V), the transistor NT2 turns off. Accordingly, the potential of the node NA becomes V_1 (V_1 is a potential that can turn on the transistor NT1), so that the transistor NT1 turns on. The potential of the node NB accordingly lowers, so that the transistor PT1 turns on because a difference in potential occurs between the source and gate thereof. Consequently, the voltage held in the power holding circuit 2 is outputted to the gate-on power supply terminal V_{gon} at a time t_2 after the power-off time.

At this time, the luminance inclination circuit 1 becomes 0 (zero) V since the supply of the gate clock signal GCLK and the luminance inclination circuit control signal XGCLK is stopped due to the cutoff of the device power supply V_{in} . Accordingly, the transistors NT3, NT4 both turn off, so that the transistor PT1 turns off since the potentials supplied to the source and gate thereof become equal to each other.

Thus, at the power-off time, it is possible to quickly change the power supply, which is fed to the gate-on power supply terminal V_{on} , from the output of the luminance inclination circuit 1 to the output of the voltage holding circuit 2 immediately after the device power supply V_{in} is cut off. Further, the gate driving circuit 4 at the power-off time asynchronously outputs the outputs of the power

holding circuit 1, which are inputted to the gate-on power supply terminal V_{gon} , to all the outputs V_{OUTi} as shown in Fig. 4 since the device power supply V_{in} fed to the all-output-on terminal V_{xon} becomes 0 (zero) V.

Consequently, all the thin film transistors in the display part 6 turn on to enable quick discharge of the residual charges in the liquid crystal. This makes it possible to clear the display, so that the afterimage can be prevented from being visible at the power-off time.

Incidentally, in the above-described embodiment, the device power supply V_{in} is directly inputted to the all-output-on terminal V_{xon} of the gate driving circuit 4, but it is also acceptable to provide a circuit that forcibly turns the voltage in the all-output-on terminal V_{xon} to low level when the device power supply V_{in} lowers to a predetermined voltage level. With such a structure, for example, even when a power supply device that does not have a capability of lowering the device power supply V_{in} in a short time is used, the potential of the all-output-on terminal V_{xon} can be forcibly lowered, so that the residual charges in the liquid crystal can be surely discharged to clear the display before the power held in the power holding circuit 2 is completely discharged.

As has been explained hitherto, according to the present invention, when the input power supply fed to the liquid crystal display device is cut off, the gate-on power supply for driving the gate lines of the display part is automatically changed from the first power supply, whose voltage is changed with time to reduce luminance inclination unevenness in the display part, to the second power supply whose voltage is kept constant for a predetermined period of time to discharge residual charges in the liquid crystal. Consequently, at the normal driving time, a good quality display image with reduced unevenness in luminance inclination can be obtained, and at the power-off time, the display due to the residual charges in the liquid crystal is cleared, so that an afterimage can be prevented from being visible. This can realize improved display quality of the liquid crystal display device both at the normal driving time and at the power-off time.

The present embodiment is to be considered in all respects as illustrative and no restrictive, and all changes which come within the meaning and range of equivalency of the claims are therefore intended to be embraced therein. The invention may be embodied in other specific forms without departing from the spirit or essential characteristics thereof.